

SECTION I—CLAIMS

Amendment to the Claims:

This listing of the claims will replace all prior versions and listings of claims in the application. Claims 30, 33-35, and 38-44 are amended herein. Claims 1-29 remain canceled herein without prejudice. No new claims are added.

Listing of Claims:

1-29. (Canceled)

30. (Currently amended) A method comprising:

receiving content for transmission from a wireless communication system having M transmit antennae and N receive antennae and N_c subcarriers, where $N_c \gg M, N$, the received content for transmission from ~~a plurality of~~ more than two of the M transmit antennae, wherein the received content is a vector of input symbols (\mathbf{s}) of size $N_c \times 1$, and wherein the N_c subcarriers is the number of subcarriers of ~~a~~ the multicarrier wireless communication channel of the wireless communication system; and

generating a rate-one, space-frequency code matrix from the received content for transmission via the ~~plurality of~~ more than two of the M transmit antennae by dividing the vector of input symbols into a number G of groups to generate subgroups and multiplying at least a subset of the subgroups by a constellation rotation precoder to produce a number G of pre-coded vectors (\mathbf{v}_g), wherein successive symbols from the same group transmitted from the same antenna are at a frequency distance that is multiples of MG subcarrier spacings, ~~wherein M represents a number of transmit antennae.~~

31. (Previously Presented) A method according to claim 30, further comprising:
dividing each of the pre-coded vectors into a number of $LM \times 1$ subvectors; and
creating an $M \times M$ diagonal matrix $D_{s_g, k} = \text{diag}\{\Theta_{M \times (k-1)+1}^T \mathbf{s}_g, \dots, \Theta_{M \times k}^T \mathbf{s}_g\}$, where $k=1 \dots L$ from
the subvectors.
32. (Previously Presented) A method according to claim 31, further comprising:
interleaving the L submatrices from the G groups to generate an $M \times Nc$ space-frequency matrix.
33. (Currently amended) A method according to claim 32, wherein the space-frequency matrix
provides MNL channel diversity, while preserving a code rate of 1 for any number of the
transmit antennae M , receive antennae ~~antenna(s)~~ N and channel tap(s) L .
34. (Currently amended) A method according to claim 30, wherein the space-frequency matrix
provides MNL channel diversity, while preserving a code rate of 1 for any number of the
transmit antennae M , receive antennae ~~antenna(s)~~ N and channel tap(s) L .
35. (Currently amended) An apparatus comprising:
a diversity agent:
to receive content for transmission from a wireless communication system having M
transmit antennae and N receive antennae and Nc subcarriers, where $Nc \gg M, N$,
the received content for transmission via a multicarrier wireless communication
channel of the wireless communication system, wherein the received content is a
vector of input symbols (\mathbf{s}) of size $Nc \times 1$, and wherein the Nc subcarriers is the
number of subcarriers of the multicarrier wireless communication channel; $[[,]]$
and
to generate a rate-one, space-frequency code matrix from the received content for
transmission on the multicarrier wireless communication channel from a ~~plurality~~

of more than two of the M transmit antennae by dividing the vector of input symbols into a number G of groups to generate subgroups and multiplying at least a subset of the subgroups by a constellation rotation precoder to produce a number G of pre-coded vectors (\mathbf{v}_g), wherein successive symbols from the same group transmitted from the same antenna are at a frequency distance that is multiples of MG subcarrier spacings,

~~, wherein M represents a number of transmit antennae.~~

36. (Previously Presented) An apparatus according to claim 35, the diversity agent further comprising:

a space-frequency encoding element, responsive to the pre-coder element, to divide each of the pre-coded vectors into a number of $LM \times 1$ subvectors, and to create an $M \times M$ diagonal matrix $D_{\mathbf{s}_g, k} = \text{diag}\{\Theta_{M \times (k-1)+1}^T \mathbf{s}_g, \dots, \Theta_{M \times k}^T \mathbf{s}_g\}$, where $k=1 \dots L$ from the subvectors.

37. (Previously Presented) An apparatus according to claim 36, wherein the space-frequency encoding element interleaves the L submatrices from the G groups to generate an $M \times Nc$ space-frequency matrix.

38. (Currently amended) An apparatus according to claim 37, wherein the space-frequency matrix provides MNL channel diversity, while preserving a code rate of 1 for any number of the transmit antennae M , receive antennae ~~antenna(s)~~ N and channel tap(s) L .

39. (Currently amended) An apparatus according to claim 35, wherein the space-frequency matrix provides MNL channel diversity, while preserving a code rate of 1 for any number of the transmit antennae M , receive antennae ~~antenna(s)~~ N and channel tap(s) L .

40. (Currently amended) A wireless communication system comprising:

a number M of omnidirectional antennas, wherein M comprises more than two omnidirectional

antennas;

a number N of receive antennae;

a number N_c of subcarriers of a multicarrier wireless communication channel of the wireless communication system, where $N_c \gg M, N$; and

a diversity agent; [[,]]

to receive content for transmission via [[a]] the multicarrier wireless communication channel, wherein the received content is a vector of input symbols (\mathbf{s}) of size $N_c \times 1$, and

~~, wherein N_c is the number of subcarriers of the multicarrier wireless communication channel, and~~

to generate a rate-one, space-frequency code matrix from the received content for transmission on the multicarrier wireless communication channel from at least a subset of the M omnidirectional antennas by dividing the vector of input symbols into a number G of groups to generate subgroups and multiplying at least a subset of the subgroups by a constellation rotation precoder to produce a number G of pre-coded vectors (\mathbf{v}_g), wherein successive symbols from the same group transmitted from the same antenna are at a frequency distance that is multiples of MG subcarrier spacings.

41. (Currently amended) A wireless communication system according to claim 40, the diversity agent further comprising:

a space-frequency encoding element, responsive to the pre-coder element, to divide each of the pre-coded vectors into a number of $LM \times 1$ subvectors, and to create an $M \times M$ diagonal

matrix $D_{\mathbf{s}_g, k} = \text{diag}\{\Theta_{M \times (k-1)+1}^T \mathbf{s}_g, \dots, \Theta_{M \times k}^T \mathbf{s}_g\}$, where $k=1 \dots L$ from the subvectors.

42. (Currently amended) A wireless communication system according to claim 41, wherein the space-frequency encoding element interleaves the L submatrices from the G groups to generate an $M \times N_c$ space-frequency matrix.
43. (Currently amended) A wireless communication system according to claim 42, wherein the space-frequency matrix provides $M N L$ channel diversity, while preserving a code rate of 1 for any number of the omnidirectional antennas M , receive antennae ~~antenna(s)~~ N and channel tap(s) L .
44. (Currently amended) A wireless communication system according to claim 40, wherein the space-frequency matrix provides $M N L$ channel diversity, while preserving a code rate of 1 for any number of the omnidirectional antennas M , receive antennae ~~antenna(s)~~ N and channel tap(s) L .